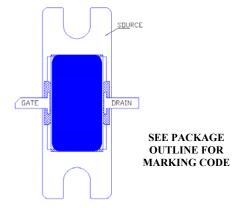


PERFORMANCE (1.8 GHz)

- ♦ 36.5 dBm Output Power (P_{1dB})
- ♦ 10.5 dB Power Gain (G_{1dB})
- ♦ 49 dBm Output IP3
- ♦ 10V Operation
- ♦ 45% Power-Added Efficiency
- ♦ Evaluation Boards Available
- ♦ Additional Design Data Available on Website
- ♦ Usable Gain to 4GHz





The FPD4000AF is a packaged depletion mode AlGaAs/InGaAs pseudomorphic High Electron Mobility Transistor (pHEMT), optimized for power applications in L-Band. The high power flangemount package has been optimized for low electrical parasitics and optimal heatsinking.

Typical applications include drivers or output stages in PCS/Cellular base station transmitter amplifiers, as well as other power applications in WLL/WLAN amplifiers.

ELECTRICAL SPECIFICATIONS AT 22°C

Parameter	Symbol	Test Conditions	Min	Тур	Max	Units		
RF SPECIFICATIONS MEASURED AT $f = 1.8$ GHz USING CW SIGNAL								
Power at 1dB Gain Compression	P_{1dB}	$V_{DS} = 10V; I_{DQ} = 720 \text{ mA}$	35.5	36.5		dBm		
		Γ_S and Γ_L tuned for Optimum IP3						
Power Gain at dB Gain Compression	G_{1dB}	$V_{DS} = 10V; I_{DQ} = 720 \text{ mA}$	9.5	10.5				
		Γ_S and Γ_L tuned for Optimum IP3						
Maximum Stable Gain	MSG	$V_{DS} = 10 \text{ V}; I_{DQ} = 720 \text{ mA}$		19		dB		
S_{21}/S_{12}		$P_{IN} = 0 dBm$, 50Ω system						
Power-Added Efficiency	PAE	$V_{DS} = 10V; I_{DQ} = 720 \text{ mA}$		45		%		
at 1dB Gain Compression		Γ_S and Γ_L tuned for Optimum IP3						
3 rd -Order Intermodulation Distortion	IP3	$V_{DS} = 10V; I_{DQ} = 720 \text{ mA}$						
		Γ_S and Γ_L tuned for Optimum IP3		-47	-44	dBc		
		$P_{OUT} = 25.5 \text{ dBm (single-tone level)}$						
Saturated Drain-Source Current	I _{DSS}	$V_{DS} = 1.3 \text{ V}; V_{GS} = 0 \text{ V}$	1.9	2.3	2.65	A		
Maximum Drain-Source Current	I_{MAX}	$V_{DS} = 1.3 \text{ V}; V_{GS} \cong +1 \text{ V}$		3.6		A		
Transconductance	G_{M}	$V_{DS} = 1.3 \text{ V}; V_{GS} = 0 \text{ V}$		2.4		S		
Gate-Source Leakage Current	I_{GSO}	$V_{GS} = -3 \text{ V}$		70	170	μΑ		
Pinch-Off Voltage	$ V_P $	$V_{DS} = 1.3 \text{ V}; I_{DS} = 8 \text{ mA}$	0.7	0.9	1.4	V		
Gate-Source Breakdown Voltage	$ V_{BDGS} $	$I_{GS} = 8 \text{ mA}$	6	8		V		
Gate-Drain Breakdown Voltage	$ V_{BDGD} $	$I_{GD} = 8 \text{ mA}$	20	22		V		
Thermal Resistivity (channel-to-case)	Θ_{CC}	See Note on following page		12		°C/W		



RECOMMENDED OPERATING BIAS CONDITIONS

Drain-Source Voltage: From 5V to 10V

Quiescent Current: From 25% I_{DSS} to 55% I_{DSS}

ABSOLUTE MAXIMUM RATINGS1

Parameter	Symbol	Test Conditions	Min	Max	Units
Drain-Source Voltage	V_{DS}	$-3V < V_{GS} < +0V$		12	V
Gate-Source Voltage	V _{GS}	$0V < V_{\rm DS} < +8V$		-3	V
Drain-Source Current	I_{DS}	For $V_{DS} > 2V$		I_{DSS}	mA
Gate Current	I_G	Forward / Reverse current		+25/-4	mA
RF Input Power ²	P _{IN}	Under any acceptable bias state		1.5	W
Channel Operating Temperature	T _{CH}	Under any acceptable bias state		175	°C
Storage Temperature	T _{STG}	Non-Operating Storage	-40	150	°C
Total Power Dissipation	P _{TOT}	See De-Rating Note below		12	W
Gain Compression	Comp.	Under any bias conditions		5	dB
Simultaneous Combination of Limits ³		2 or more Max. Limits		80	%

 $^{{}^{1}}T_{Ambient} = 22^{\circ}C$ unless otherwise noted ²Max. RF Input Limit must be further limited if input VSWR > 2.5:1

Notes:

Operating conditions that exceed the Absolute Maximum Ratings will result in permanent damage to the device.

Total Power Dissipation defined as: $P_{TOT} = (P_{DC} + P_{IN}) - P_{OUT}$, where:

P_{DC}: DC Bias Power P_{IN}: RF Input Power P_{OUT}: RF Output Power

Total Power Dissipation to be de-rated as follows above 22°C:

 $P_{TOT} = 12 - (0.083 \text{W/}^{\circ}\text{C}) \times T_{PACK}$

where T_{PACK} = source tab lead temperature above 22 °C (coefficient of de-rating formula is the Thermal Conductivity)

Example: For a 55°C source lead temperature: $P_{TOT} = 12 - (0.083 \times (55 - 22)) = 9.3W$

Note on Thermal Resistivity: The nominal value of 12°C/W is measured with the package mounted on a large heatsink with thermal compound to ensure adequate contact. The package temperature is referred to the Source flange.

HANDLING PRECAUTIONS

To avoid damage to the devices care should be exercised during handling. Proper Electrostatic Discharge (ESD) precautions should be observed at all stages of storage, handling, assembly, and testing. These devices should be treated as Class 1A per ESD-STM5.1-1998, Human Body Model. Further information on ESD control measures can be found in MIL-STD-1686 and MIL-HDBK-263.

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³Users should avoid exceeding 80% of 2 or more Limits simultaneously



BIASING GUIDELINES

- Active bias circuits provide good performance stabilization over variations of operating temperature, but require a larger number of components compared to self-bias or dual-biased. Such circuits should include provisions to ensure that Gate bias is applied before Drain bias, otherwise the pHEMT may be induced to self-oscillate. Contact your Sales Representative for additional information.
- ➤ Dual-bias circuits are relatively simple to implement, but will require a regulated negative voltage supply for depletion-mode devices such as the FPD4000AF.
- Self-biased circuits employ an RF-bypassed Source resistor to provide the negative Gate-Source bias voltage, and such circuits provide some temperature stabilization for the device. A nominal value for circuit development is $0.7~\Omega$ for the recommended 720mA operating point. This approach will require a DC Source resistor capable of at least 365mW dissipation.
- ➤ The recommended 720mA bias point is nominally a Class AB mode. A small amount of RF gain expansion prior to the onset of compression is normal for this operating point.

PACKAGE OUTLINE

(dimensions in millimeters – mm)

PACKAGE MARKING CODE

Example:

*f*1ZD P2F

f = Filtronic

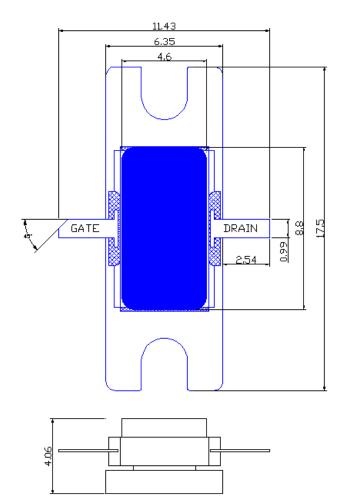
1ZD = Lot and Date Code

P2F = Status, Part Code, Part Type

Status: D=Development P = Production

Part Code denotes model (e.g. FPD4000AF)

Part Type: F = FET (pHEMT)



All information and specifications subject to change without notice.

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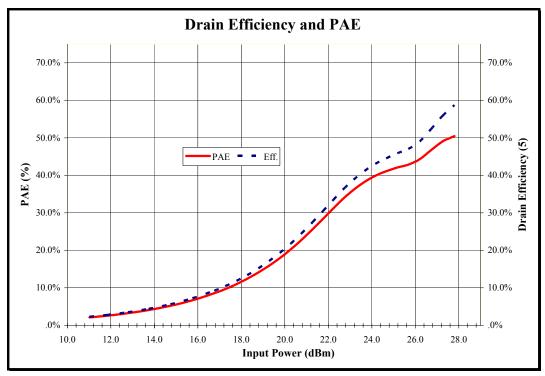
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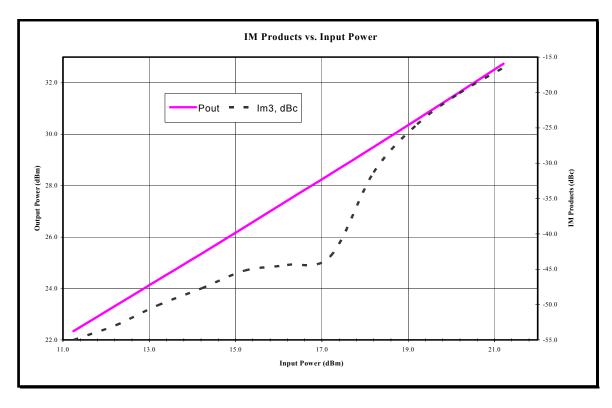


• TYPICAL RF PERFORMANCE ($V_{DS} = 10V I_{DQ} = 720 \text{ mA}$ f = 2000 MHz):

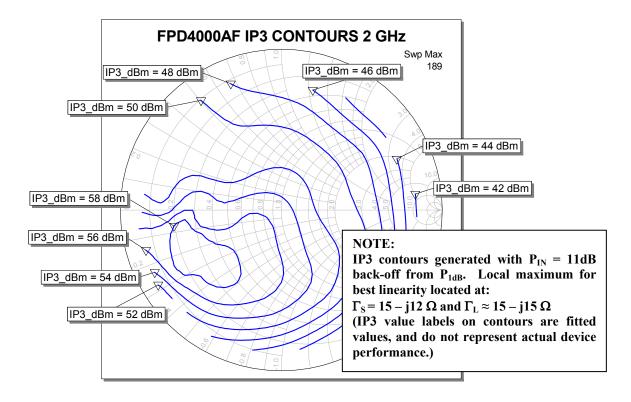








Note: Graph above shows Input and Output power as single carrier or single-tone levels.



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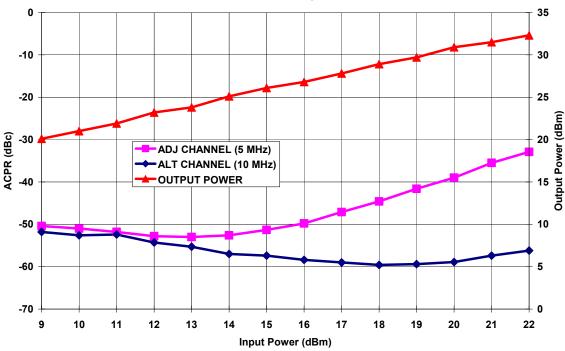
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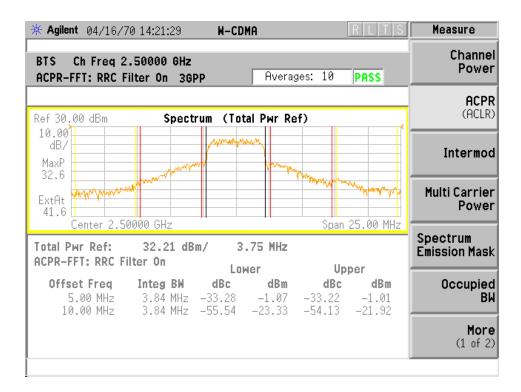
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ACPR measurement at 4 dB back-off from P_{1dB} with WCDMA BTS Forward modulation:

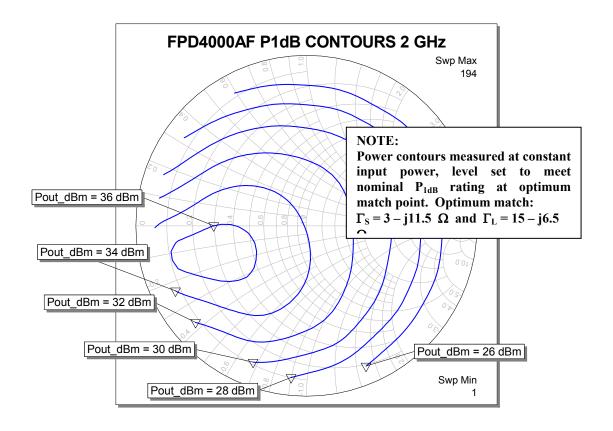


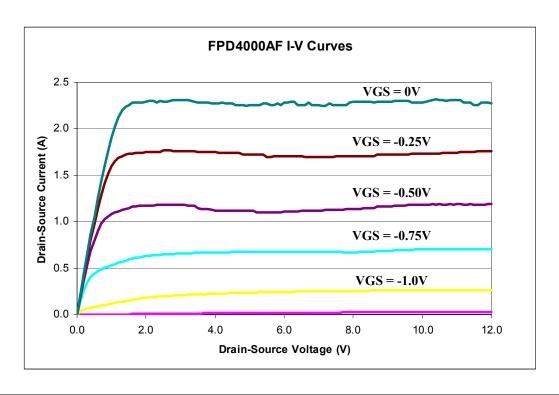
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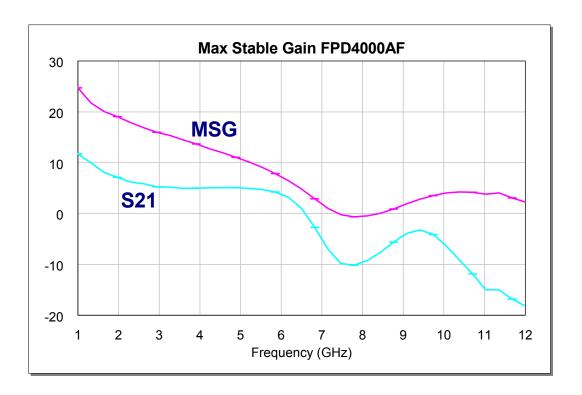
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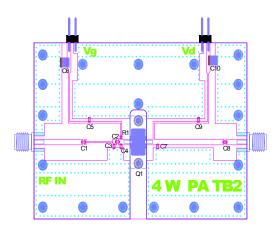
RF PERFORMANCE OVER FREQUENCY:



Note: The FPD4000AF is suitable for applications up to 4 GHz.



STANDARD EVALUATION BOARD (1.70–1.85 GHZ):



NOTE: AutoCADTM drawing available on Website

Bill of Materials for Evaluation Board, FPD4000AF EV-BL-000026-002-A						
Designator	Part Number	Description	Quantity			
C1	ATC600S0R9JW250	Capacitor, 0.9 pF, 0603, ATC 600, tol. <u>+</u> 5%	1			
C2	ATC600S3R0JW250	Capacitor, 3.0 pF, 0603, ATC 600, tol. <u>+</u> 5%	1			
C3	ATC600S1R8JW250	Capacitor, 1.8 pF, 0603, ATC 600, tol. <u>+</u> 5%	1			
C4	ATC600S1R0JW250	Capacitor, 1.0 pF, 0603, ATC 600, tol. <u>+</u> 5%	1			
C5	ATC600S330JW250	Capacitor, 33 pF, 0603, ATC 600, tol. <u>+</u> 5%	1			
C6	T491B105M035AS7015	Capacitor, 1 μF, SMD-B, Kemet, tol. ±20%	1			
C7	ATC600S1R5JW250	Capacitor, 1.5 pF, 0603, ATC 600, tol. <u>+</u> 5%	1			
C8	ATC600S330JW250	Capacitor, 33 pF, 0603, ATC 600, tol. <u>+</u> 5%	1			
C9	ATC600S330JW250	Capacitor, 33 pF, 0603, ATC 600, tol. <u>+</u> 5%	1			
C10	T491B105M035AS7015	Capacitor, 1 μF, SMD-B, Kemet, tol. ±20%	1			
R1	RCI-0603-10R1J	Resistor, 100 Ω , 0603, IMS, tol. \pm 5%	1			
Q1	FPD4000AF	Packaged Discrete pHEMT, Filtronic	1			
	PC-SP-000022-002	PCB, FPD4000AF Eval Board, 2 GHz	1			
	142-0711-841	Connector, RF, SMA End Launch, Jack Assy, Johnson	2			
	AMP-103185-2	Connector, DC, 0.100 on center, 0.025 sq. posts, Tyco				
	TF-SP-000025	Test Fixture Base, Flange Mount Package, 2 GHz	1			
		Screw, #2-56	20			